



MAX PLANCK INSTITUTE
FOR DYNAMICS OF COMPLEX
TECHNICAL SYSTEMS
MAGDEBURG



COMPUTATIONAL METHODS IN
SYSTEMS AND CONTROL THEORY

Numerical Methods in Control and Optimization of Dynamical Systems

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BIMoS Days at TU Berlin

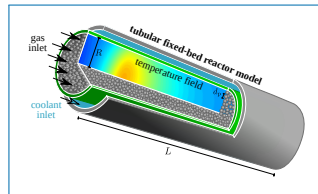
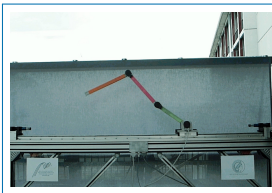
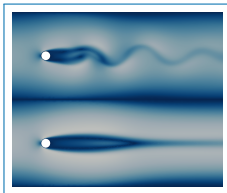
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Control



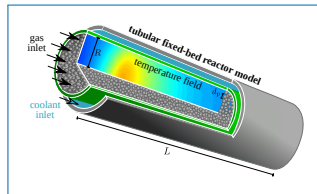
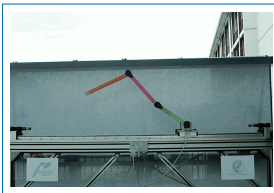
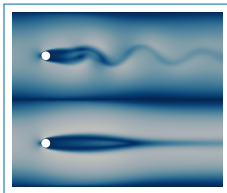
Optimization

- Stabilization of
- Small Deviations
- Linear Theory
- General approaches

- Large changes
- Nonlinear methods
- Data-driven approaches



Control



Optimization

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Optimal Control

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1. Introduction to Linear Time Invariant Systems
2. Linear Approximations vs. Nonlinear Control
3. Introduction to Robust Control
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Typical Situation



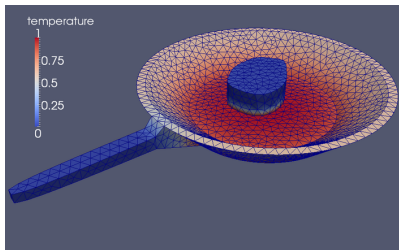
- Fry a steak
- The cook controls the heat at the fireplace
- and observes the process, e.g. via measuring the temperature in the inner



- The model

$$\begin{aligned}\dot{\theta} &= \nabla \cdot (\nu \nabla \theta) && \text{in } (0, \infty) \times \Omega, \\ \theta &= u, && \text{at the hob,} \\ \theta(0) &= 0.\end{aligned}$$

- The cook controls the heat at the fireplace, which we denote by u
- and observes the process, e.g. he measures the temperature y in the center: $y = f(\theta)$.



- The model:

$$\begin{aligned}\dot{\theta} &= \nabla \cdot (\nu \nabla \theta), \\ \theta &= u, & (\text{at the hob}), \\ \theta(0) &= 0.\end{aligned}$$

- The cook controls the heating u
- and observes the process via $y = f(\theta)$.

- A *Finite Element* discretization [GAUL'13] of the problem leads to the finite dimensional model

$$\begin{aligned}\dot{\theta}(t) &= A\theta(t) + Bu(t), \quad \theta(0) = 0, \\ y(t) &= C\theta(t),\end{aligned}$$

a linear time invariant (LTI) system.



Linear State Space System

$$\begin{aligned}\dot{x}(t) &= Ax(t) + Bu(t), & x(0) &= x_0, \\ y(t) &= Cx(t) + Du(t),\end{aligned}$$

with

- $x(t) \in \mathbb{R}^n$: the system's state
- $u(t) \in \mathbb{R}^m$: the input or control
- $y(t) \in \mathbb{R}^q$: the output or measurements
- $n, m, q \in \mathbb{N}$: the system dimensions



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- $A \in \mathbb{R}^{n \times n}$: the system matrix
- $B \in \mathbb{R}^{n \times m}$: the input matrix
- $C \in \mathbb{R}^{q \times n}$: the output matrix
- $D \in \mathbb{R}^{q \times m}$: the throughput



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Definition (Exponential Stability and Stabilizability)

- A (possibly nonlinear) dynamical system can be called exponentially stable, if all solutions x (starting in a neighborhood of the origin), decay to the origin exponentially, i.e.

$$\|x(t)\| \leq M e^{-\lambda t} \|x(0)\|, \quad t > 0,$$

for some constants $M, \lambda > 0$.



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- The LTI system $\dot{x} = Ax + Bu$ is called stable, if

$$\|e^{tA}\| \leq Me^{-\lambda t}, \quad t > 0.$$

- The LTI system is called stabilizable, if there exists $K \in \mathbb{R}^{m \times n}$ such that

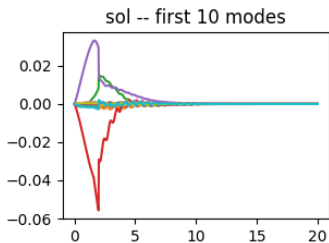
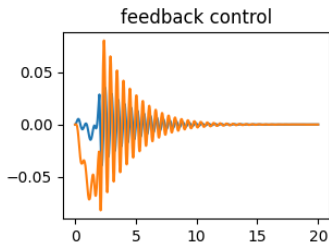
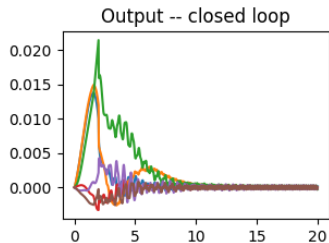
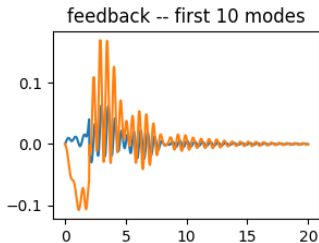
$$\|e^{t(A-BK)}\| \leq Me^{-\lambda t}, \quad t > 0.$$



Exponential Stability

$$\|x(t)\| \leq Me^{-\lambda t} \|x(0)\|, \quad t > 0,$$

- $M > 1$ captures transient behavior
- λ denotes the rate of decay.





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Most of my research – bridging the gap

Navier-Stokes Equations:

$$\dot{v} + (v \cdot \nabla)v - \nu \Delta v + \nabla p = \mathcal{B}u,$$
$$\nabla \cdot v = 0$$

controlled
by

LTIs:

$$\dot{x}(t) = Ax(t) + Bu(t), \quad x(0) = x_0,$$
$$y(t) = Cx(t) + Du(t),$$

- nonlinear
- ∞ -dimensional
- accurate



- not nonlinear
- not ∞ -dimensional
- not accurate



Linear vs. Nonlinear

How can linear theory work for nonlinear systems?

The self-referential promise of linear control theory

the linear controller works



everything is under control



Set point stabilization stabilization of a nonlinear system:

$$\dot{x} = f(x) + Bu$$

Linearization about $x_* = 0$ with $f(x_*) = 0$:

$$\dot{x} = f(0) + D_x f(0)x + r(x) + Bu = A_*x + Bu + r(x)$$

with

- $A_* = D_x f(0)$ – the Jacobian of f at 0
- $r(\cdot) \in o(\|\cdot\|)$, i.e. $\lim_{x \rightarrow 0} \frac{\|r(x)\|}{\|x\|} = 0$



Theorem

If the linearized (around $x_* = 0$) problem can be stabilized, then the nonlinear problem can be stabilized locally around $x_* = 0$.

Proof:

- Let $A_* - BK$ be stable, with $e^{t(A_* - BK)} \leq Me^{-\lambda t}$, and set $u = -Kx$.
- By the *Variation of Constant* formula we have

$$x(t) = e^{t(A_* - BK)}x(0) + \int_0^t e^{(t-s)(A_* - BK)}r(x(s)) \, ds.$$

- By $r \in o$, for any $\eta > 0$, there exists $\delta(\eta) > 0$ such that $\|r(x)\| < \eta\|x\|$, for $\|x\| < \delta(\eta)$
- and with *Gronwall inequality*, we arrive at the inequality

$$\|x(t)\| \leq \|x(0)\| Me^{-(\lambda - \eta M)t}$$

that holds for $t > 0 \dots$

\dots as long as $\|x(t)\| < \delta(\eta)$.



$$\|x(t)\| \leq \|x(0)\| M e^{-(\lambda - \eta M)t}$$

that holds for $t > 0$...

... as long as $\|x(t)\| < \delta(\eta)$.

Now

1. by continuity of x , for $x(0) < \delta(\eta)$, we have $x(0 + \varepsilon) < \delta(\eta)$ – **this inequality is not void!**
2. choose $\eta < \frac{\lambda}{M}$, then $\lambda - \eta M > 0$ and $e^{-(\lambda - \eta M)t} < 1$ and exponentially decaying
3. choose $x(0)$, with $\|x(0)\| < \frac{\delta(\eta)}{M}$ (note that $M > 1$), so that

$$\|x(t)\| \leq \|M\| \|x(0)\| e^{-(\lambda - \eta M)t} < \delta(\eta) e^{-(\lambda - \eta M)t}$$

for $t > 0$.



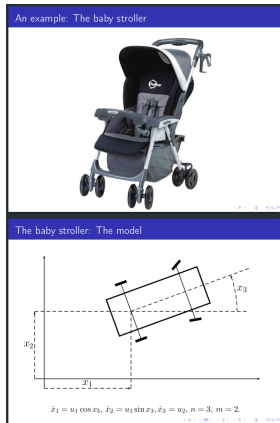


- Linearized stability sufficient for nonlinear (local) stabilizability
- Similar things hold for ∞ -dimensional systems (aka control of nonlinear PDEs); [RAYMOND'06, BREITEN&KUNISCH'14].
- What if there is no linearization?
 - In finite dimensions – f not even Lipschitz – locally nonunique solutions.
 - Existence of a suitable linearization is a general problem in ∞ -dimensions.



Some side remarks

- Linearized stability sufficient for nonlinear (local) stabilizability
- Similar things hold for ∞ -dimensional systems (aka control of nonlinear PDEs); [RAYMOND'06, BREITEN&KUNISCH'14].
- What if there is no linearization?
 - In finite dimensions – f not even Lipschitz – locally nonunique solutions.
 - Existence of a suitable linearization is a general problem in ∞ -dimensions.
- Sometimes, better stay nonlinear: A baby stroller is (nonlinearly) controllable but not linearly stabilizable.



Lecture by [J.M. CORON'11]



So, with a linearization A_* and with a controller K , so that the LTI

$$\dot{x} = (A_* - BK)x$$

is stable, the nonlinear system

$$\dot{x} = f(x) - BKx$$

is locally stabilized around x_* .

Q: What if A_* , i.e. the linearization, is faulty?

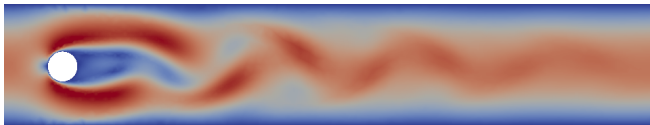
Q: How to compute K ?



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Flow Control Problem

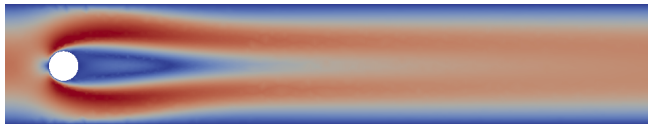


Feedback Control

Problem: The steady state is unstable: any perturbation – no matter how small – will trigger a transition into a periodic regime.

Goal: Stabilizing feedback controller that can handle:

- limited measurements,
- system uncertainties.





Flow Control Problem

Idea: Linearization-based feedback control for stabilization of the steady state.

[RAYMOND06, BENNER&JH'15, BREITEN&KUNISCH'14]

$$\begin{aligned}\dot{v} + (v \cdot \nabla)v - \nu \Delta v + \nabla p &= Bu, \\ \nabla \cdot v &= 0\end{aligned}$$

Linearization &
Semi-Discretization

$$\begin{aligned}\dot{v} - Av - J^\top p &= Bu, \\ Jv &= 0\end{aligned}$$



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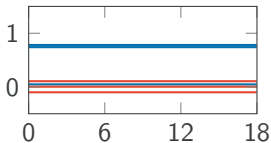
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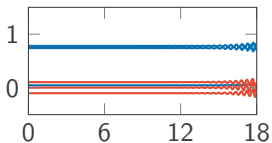
Fragility of Observer-Based Controllers

LQG controllers have no guaranteed robustness margins and will likely fail in the presence of system uncertainties.

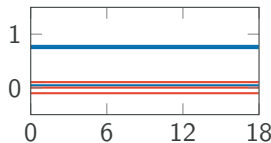
LQG-feedback



corrupted LQG-feedback



corrupted state-feedback





In fact: [IEEE TRANSACTIONS ON AUTOMATIC CONTROL ('78)]:

Guaranteed Margins for LQG Regulators

JOHN C. DOYLE

Abstract—There are none.

Good news: \mathcal{H}_∞ controllers work under uncertainties like

- [CURTAIN'03]: Galerkin approximations of evolution systems,
- [BENNER&JH'17]: stable mixed-FEM approximation of the flow equations,
- [BENNER&JH'16]: *errors in the linearization point*,
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that can be qualified as a coprime factor perturbations.

Moreover,

- [THIS TALK, JH'21]: the coprime factor perturbation depends smoothly on the linearization error.
- [THIS TALK, BENNER&JH&WERNER'21]: we can compute the controller and its robustness



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Transfer functions

Mapping of inputs (controls) to outputs (measurements) in frequency domain, i.e., after Laplace transform of the system.

$$\begin{array}{lcl} \dot{x} = Ax + Bu & \xrightarrow{\mathcal{L}(s)} & sX(s) = AX(s) + BU(s) \\ y = Cx & & Y(s) = CX(s) \end{array}$$



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1. A *nominal* system has the transfer function

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1. A *nominal* system has the transfer function

$$G(s) = C(sI - A)^{-1}B \in \mathbb{C}^{q,r}.$$

2. But uncertainty in the operator gives another transfer function

$$G_{\Delta}(s) = C(sI - A - \delta_A)^{-1}B \in \mathbb{C}^{q,r}.$$



Coprime Factorization

Given a transfer function $G(s)$ of a linear system,

$$G(s) = M^{-1}(s)N(s)$$

is a **(left) coprime factorization** if there exist $X(s)$, $Y(s)$ such that the Bezout identity

$$M(s)X(s) + N(s)Y(s) = I$$

holds. Here, N , M , X , Y are all rational matrix functions with all poles in the open left half of the complex plane, i.e., they all represent stable linear systems.



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Coprime Factor Perturbation

$$G_{\Delta}(s) = [N(s) + \delta_N(s)] [M(s) + \delta_M(s)]^{-1}(s) \approx G(s) = N(s)M^{-1}(s),$$

where $N + \delta_N, M + \delta_N$ are stable.



Next we will show that

- Inexact linearizations of incompressible Navier-Stokes equations
- can be qualified as a coprime factor uncertainty
- that smoothly depends on the linearization error.

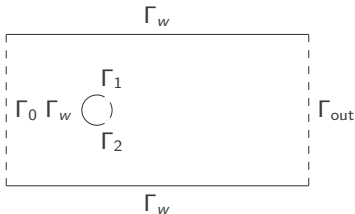
So that the standard \mathcal{H}_∞ -theory for robust controller design applies.



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We consider



where

- V ... velocity,
- P ... pressure,
- ν ... diffusion parameter,

$$\begin{aligned}\dot{V} + (V \cdot \nabla)V + \nabla P - \nu \Delta V &= 0, \\ \operatorname{div} V &= 0, \quad \text{in } \Omega,\end{aligned}$$

$$\begin{aligned}\nu \frac{\partial V}{\partial n} - nP &= 0 \text{ on } \Gamma_{\text{out}}, \\ V &= 0 \text{ on } \Gamma_w, \\ V &= ng_0 \cdot \alpha \text{ on } \Gamma_0, \\ V &= ng_1 \cdot u_1 \text{ on } \Gamma_1, \\ V &= ng_2 \cdot u_2 \text{ on } \Gamma_2,\end{aligned}$$

- g_0, g_1, g_2 ... spatial shape functions,
- u_1, u_2 ... scalar input functions,
- α ... magnitude of the inflow velocity,
- n ... normal vector at the boundaries.



A linearized I/O model is obtained as follows:

1. We relax the Dirichlet control $V|_{\Gamma_1} = ng_1 u - \epsilon(\nu \frac{\partial V}{\partial n} - Pn)$
2. Let v_α be the steady state solution for zero inputs, and let $v_\delta(t) = V(t) - v_\alpha$ the deviation.
3. We consider the linearization

$$\dot{v}_\delta + (v_\delta \cdot \nabla)v_\alpha + (v_\alpha \cdot \nabla)v_\delta + \nabla p_\delta - \nu \Delta v_\delta = 0$$

that is a valid approximation as long as v_δ is small.



Then, with

$$\mathcal{H}_{div} := \{v \in L^2(\Omega) : \operatorname{div} v = 0, v \cdot n = 0 \text{ on } \Gamma_w \cap \Gamma_{out}\}$$

as the state space, the (orthogonal) *Leray*-projector

$$\Pi \in \mathcal{L}(L^2(\Omega)) : L^2(\Omega) \mapsto \mathcal{H}_{div},$$

and $x := \Pi v_\delta$ the model reads¹

$$\begin{aligned}\dot{x} &= A_\alpha x + \Pi B u \quad \text{in } \mathcal{H}_{div}, \\ y &= C x\end{aligned}$$

where

- $A_\alpha : \mathcal{D}(A_\alpha) \subset \mathcal{H}_{div} \rightarrow \mathcal{H}_{div}$ is the *Oseen* operator
- $\Pi B : \mathbb{R}^2 \rightarrow \mathcal{H}_{div}$ is the input operator
- $C : \mathcal{H}_{div} \rightarrow \mathbb{R}^q$ is the output operator

¹The pressure p_δ is gone, since Π maps along the orthogonal complement of the gradient



Boundedness of the input operator

Lemma (JH'21, Benner&JH'18)

If $g_i \in H_{00}^{1/2}(\Gamma_i)^2$, $i = 1, 2$, and $\varepsilon > 0$, then the input operator $B: \mathbb{R}^2 \rightarrow L^2(\Omega)$ for the Oseen system that realizes

$$V = ng_i \mathbf{u}_i - \varepsilon \left(\nu \frac{\partial V}{\partial n} - nP \right) \quad \text{on } \Gamma_i, \quad i = 1, 2$$

is bounded.

Outline of the proof:

- By definition $B = \Pi B$, with Π being the orthogonal projector onto \mathcal{H}_{div} .
- We show that $\langle \Pi B \mathbf{u}, w \rangle_{L^2(\Omega)} = \langle B \mathbf{u}, \Pi w \rangle_{L^2(\Omega)}$.
- Thus, $\langle B \mathbf{u}, w \rangle_{L^2(\Omega)} = -\frac{1}{\varepsilon} \sum_{i=1,2} \int_{\Gamma_i} \Pi w \cdot (g_i n) \, ds \, \mathbf{u}_i$.
- Since $\Pi w \cdot n \in H^{-1/2}(\Gamma_i)$, it follows that $\Pi B: \mathbb{R}^2 \rightarrow L^2(\Omega)$ is bounded.
- By definition $\Pi B = B$.

${}^2H_{00}^{1/2}(\Gamma_i)$ contains those functions out of $H^{1/2}(\Gamma_i)$ that are boundedly extendable by 0 to the complete boundary.



- ✓ The linearized model is a standard (A, B, C) system
 - we know: A_α is the generator of a C_0 -semi group [RAYMOND'06]
 - we choose: C to be bounded
 - we have just shown: ΠB is bounded.

→ The theory for robust stabilization of linearization errors applies.

← Assume that the linearization point v_α is uncertain

- that is $v_\alpha \leftarrow v_\alpha + \delta_v$
- then A is perturbed $A \leftarrow A + \delta_A$
- as is the transferfunction

$$G_\delta(s) = C(sI - A - \delta_A)^{-1}B$$



Theorem (JH'21)

Consider the perturbed Oseen system and let $L \in \mathcal{L}(\mathbb{R}^k, V^0)$ and $\delta_A(\delta_v)$ be such that $(A + \delta_A - LC)$ is exponentially stable for all δ_A small. Then the associated transferfunction G_δ has a coprime factorization

$$G_\delta = [N + \delta_N][M + \delta_M]^{-1},$$

where $NM^{-1} = G$ is the transferfunction associated with the unperturbed system, and

$$\|\delta_N\|_{\mathcal{H}_\infty} \rightarrow 0 \quad \text{and} \quad \|\delta_M\|_{\mathcal{H}_\infty} \rightarrow 0$$

as $\delta_v \rightarrow 0$.



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$$G_\delta = [N + \delta_N][M + \delta_M]^{-1},$$

where $NM^{-1} = G$ is the transferfunction associated with the unperturbed system, and

$$\|\delta_N\|_{\mathcal{H}_\infty} \rightarrow 0 \quad \text{and} \quad \|\delta_M\|_{\mathcal{H}_\infty} \rightarrow 0$$

as $\delta_V \rightarrow 0$.

Remark on the existence of L

The existence of the uniformly stabilizing L (as we have assumed it here) is much less critical than the existence of a robust controller (because the L is a *state feedback*).



1. The perturbation δ_N has the representation³

$$\delta_N(s) = C\delta_A(sI - A + LC)^{-1}(sI - A - \delta_A + LC)^{-1}\Pi B,$$

2. and can be realized as a cascaded system

$$\dot{v}_1 = (A + \delta_A - LC)v_1 + \Pi Bu, \quad (\mathcal{F}_1)$$

$$\dot{v}_2 = (A - LC)v_2 + v_1 \quad (\mathcal{F}_2)$$

$$y = C\delta_A v_2,$$

in the time domain.

3. This results in the transferfunction (in the time domain):

$$y = C\delta_A \mathcal{F}_2 \mathcal{F}_1 u.$$



For the transfer function in the time domain

$$y = C\delta_A \mathcal{F}_2 \mathcal{F}_1 u$$

we have that:

1. Certainly $\|C\delta_A\| \rightarrow 0$ if $\|\delta_A\| \rightarrow 0$, but only on function spaces with sufficient regularity. (The operator δ_A contains spatial derivatives)
2. Therefore, we use
 - the uniform stability of $A + \delta_A - LC$
 - and the analyticity of the semi-group that is generated by $A - LC$to show that $\mathcal{F}_2 \mathcal{F}_1$ provides the needed regularity.



3. By means of a classical result⁴, that connects frequency- and time domain, we infer that dass

$$\|\delta_N\|_{\mathcal{H}_\infty} \leq \|C\delta_A\mathcal{F}_2\mathcal{F}_1\|_{L^2 \rightarrow L^2},$$

so that $\|\delta_A\| \rightarrow 0$ implies that

$$\|\delta_N\|_{\mathcal{H}_\infty} \rightarrow 0.$$



³Benner&JH(2016) *IFAC PapersOnLine* based on the textbook by Curtain&Zwart(1995)

⁴Weiss(1991) *Representation of shift-invariant operators on L^2 by H^∞ transfer functions*



How can this help?

One can show [MUSTAFA&GLOVER'91, BENNER/JH/WERNER'19, ZHOU/DOYLE/GLOVER'96]:

- that uncertainty in the linearization can be formulated as an *normalized \mathcal{H}_∞ robust control problem*
- and that \mathcal{H}_∞ robust controller K of robustness margin γ will stabilize the perturbed system if

$$\|[\delta_N \quad \delta_M]\|_{\mathcal{H}_\infty} < \gamma^{-1}.$$

So let's compute such a controller...



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\mathcal{H}_∞ Riccati Equations

[ZHOU/DOYLE/GLOVER'95]

Given some simplifying assumptions, there exists an admissible controller $K(s) \iff$:

1. There exists a stabilizing solution $X_\infty = X_\infty^\top \geq 0$ to the regulator Riccati equation

$$A^\top X_\infty + X_\infty A + C^\top C - (1 - \gamma^{-2}) X_\infty B B^\top X_\infty = 0.$$

2. There exists a stabilizing solution $Y_\infty = Y_\infty^\top \geq 0$ to the filter Riccati equation

$$A Y_\infty + Y_\infty A^\top + B B^\top - (1 - \gamma^{-2}) Y_\infty C^\top C Y_\infty = 0.$$

3. It holds $\gamma^2 > \lambda_{\max}(Y_\infty X_\infty)$.



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The central (or minimum entropy) controller $\hat{K}(s) = \hat{C}(sI - \hat{A})^{-1}\hat{B}$ is given by

$$\hat{A} = A - (1 - \gamma^{-2}) B B^\top X_\infty - Z_\infty Y_\infty C^\top C, \quad \hat{B} = Z_\infty Y_\infty C^\top, \quad \hat{C} = -B^\top X_\infty,$$

with $Z_\infty = (I_n - \gamma^{-2} X_\infty Y_\infty)^{-1}$.



Arising Challenges

Large-Scale Matrix Equations

How to solve the arising large-scale sparse Riccati equations

$$A^\top X_\infty + X_\infty A + C_1^\top C_1 + X_\infty (\gamma^{-2} B_1 B_1^\top - B_2 B_2^\top) X_\infty = 0?$$

- **Low-rank Riccati iteration** solves indefinite Riccati equations by an approximation

$$X_\infty \approx ZZ^\top,$$

with $Z \in \mathbb{R}^{n \times r}$ and $r \ll n$.

[LANZON/FENG/ANDERSON '07, BENNER/HEILAND/W. '22A]



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High Dimensional Controller

How to construct a low dimensional $\hat{K}(s)$ for faster evaluation?

- Use model order reduction based on X_∞ and Y_∞ .



(Still) Stabilizing Reduced-Order Controller

Notation:

- normalized left coprime factorizations $G = M^{-1}N$, $G_r = M_r^{-1}N_r$
(for computation see [BENNER/HEILAND/W. '19]),
- $\beta = \sqrt{1 - \gamma^{-2}}$.

The approximation error of the \mathcal{H}_∞ balanced truncation is given by

$$\|[\beta(N - N_r) \quad M - M_r]\|_{\mathcal{H}_\infty} =: \beta\hat{\epsilon} \leq \beta\epsilon = 2 \sum_{k=r+1}^n \frac{\beta\sigma_k^{\mathcal{H}_\infty}}{\sqrt{1 + \beta^2 (\sigma_k^{\mathcal{H}_\infty})^2}},$$

where $\sigma_k^{\mathcal{H}_\infty}$ are the \mathcal{H}_∞ *characteristic values*.



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where $\sigma_k^{\mathcal{H}_\infty}$ are the \mathcal{H}_∞ *characteristic values*.

Theorem

[MUSTAFA/GLOVER '91]

The reduced-order \mathcal{H}_∞ controller is guaranteed to stabilize the full-order system if

$$\hat{\epsilon}(\beta + \gamma) < 1 \quad \text{or} \quad \epsilon(\beta + \gamma) < 1.$$



Numerical Realization of the DAE Structure

For consistent initial values, i.e., $Jv_0 = 0$, the semi-discretized Navier-Stokes equation can be realized by an ODE system:

$$\begin{aligned} E\dot{v} &= Av + J^\top p + Bu, \\ 0 &= Jv, \\ y &= Cv, \end{aligned}$$



$$\begin{aligned} E\dot{v} &= \Pi^\top A \Pi v + \Pi^\top B, \\ y &= C \Pi v, \end{aligned}$$

where $\Pi = I_{n_v} - E^{-1}J^\top(JE^{-1}J^\top)^{-1}J$ is the discrete Leray projection.



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Implicit Realization

[HEINKENSCHLOSS&SORENSEN&SUN '08]

The explicit projection Π can be avoided in the numerical methods by solving saddle point problems of the type

$$\begin{bmatrix} A + s_i E & J^\top \\ J & 0 \end{bmatrix} \begin{bmatrix} X \\ * \end{bmatrix} = \begin{bmatrix} Y \\ 0 \end{bmatrix}.$$



Linearization Uncertainties

In general, an uncertainty \mathcal{A}_Δ in the linearization \mathcal{A} ...

$$\begin{aligned}\mathcal{E}\dot{x}(t) &= \mathcal{A}x(t) + \mathcal{B}u(t), \\ y(t) &= \mathcal{C}x(t)\end{aligned}$$



$$\begin{aligned}\mathcal{E}\dot{x}(t) &= [\mathcal{A} + \mathcal{A}_\Delta]x(t) + \mathcal{B}u(t), \\ y(t) &= \mathcal{C}x(t)\end{aligned}$$

... is an additive uncertainty in the transfer function

$$G(s) = \mathcal{C}(s\mathcal{E} - \mathcal{A})^{-1}\mathcal{B}$$



$$\begin{aligned}G_\Delta(s) &= \mathcal{C}(s\mathcal{E} - \mathcal{A} - \mathcal{A}_\Delta)^{-1}\mathcal{B} \\ &= G(s) + \tilde{G}(s)\end{aligned}$$

where $\tilde{G}(s) = \mathcal{C}\mathcal{A}_\Delta(s\mathcal{E} - \mathcal{A})^{-1}(s\mathcal{E} - \mathcal{A} - \mathcal{A}_\Delta)^{-1}\mathcal{B}$.



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Additive uncertainties can be compensated by robust \mathcal{H}_∞ controller design.

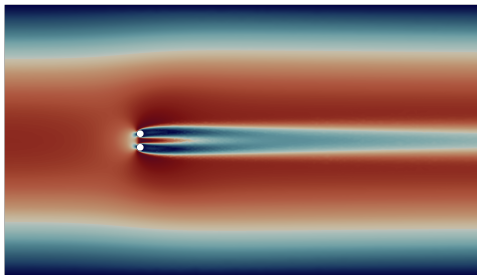


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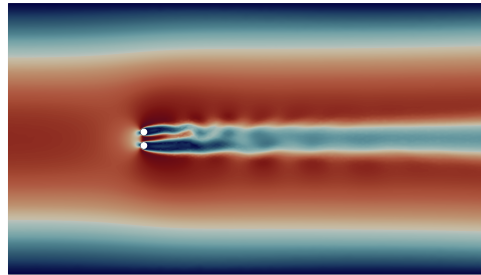


Setup: Stabilization of the steady state

[BENNER&JH&WERNER'21]



(a) Steady state.

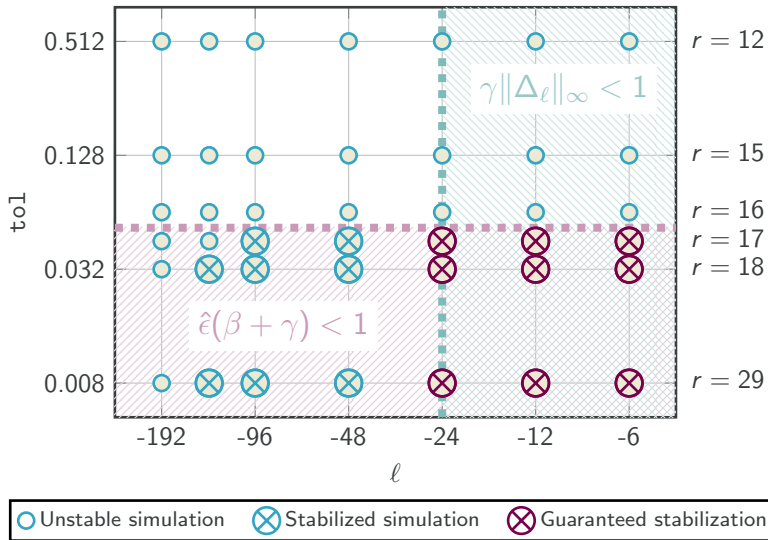


(b) Natural flow.

- Navier-Stokes equations discretized by Taylor-Hood finite elements
- system order $n = 51\,337$
- Reynolds number 60
- boundary control: individual rotation of both cylinders
- observations: 3 velocity sensors in the wake behind the cylinders



Double-cylinder: Results [BENNER/HEILAND/W. '22]





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